It is introduced by W.R Evans in 1948 for the analysis of control systems.

It is a powerful tool for adjusting the location of closed loop poles to achieve the desired system performance by varying one or more system parameters.

It is a powerful method of analysis and design for stability and transient response.

Typical Sketches of Root Locus Plots

Locus on Real Axis





TYPICAL SKETCHES OF ROOT LOCUS PLOTS





Step 1:-

Locate the Poles and Zeros of G(s).H(s) on the S-plane.
 The Root locus branch start from open loop poles and terminate at Zeros.
 n – number of poles, m – number of zeros

Step 2:-

Determine the Root locus on real axis.ie, Take a test point on real axis.

Step 3:-

> Determine the asymptotes of Root locus branches and meeting point of asymptotes with real axis and Centroid.

i)Angle of Asymptote = $[\pm 180^{\circ}(2q+1)]/(n-m)$, q=0,1,2,3...(n-m)

ii) Centroid= [sum of poles-sum of zeros]/[n-m]

Step 4:-

Find the break-away and break-in points.
i) If there is a Root locus on real axis between
2 poles then there exist a break-away point

ii) If there is a Root locus on real axis between2 zeros then there exist a break-in point

iii) If there is a Root locus on real axis between poles and zeros then there may be or may not be break-away or break —in point

Step 5:-

➢ If there is <u>complex pole</u> then determine the angle of departure from the complex pole.

> Angle of Departure = [180°-(sum of angles of vector to the complex pole A from other poles) + (sum of angles of vector to the complex pole A from other zeros)]

If there is <u>complex zero</u> then determine the angle of arrival at the complex zero.

Angle of Departure= [180°-(sum of angles of
vector to the complex zero
A from all other zeros) +
(sum of angles of vector to
the complex zero A from
other poles)]

Step 6:-

Find the points where the Root loci may cross the imaginary axis.

Step 7:-

➤ Take a series of test point in the broad neighborhood of the origin of the S-plane and adjust the test point to satisfy angle criterion.

Sketch the Root locus by joining the test point by smooth curve.

If the total no. of poles and zeros on real axis to the right of test point is

i)Odd no. - Test point lies on real axisii) Even no. - Test point does not lie on real axis

>Step 8:-The value of gain K at any point on the locus can be determined from Magnitude condition. The magnitude condition is given by Gain K at a point = [Product of length of (S=Sa)vectors from poles to the point S=Sa]/ **[Product of length of** vectors from finite zeros to the point S=Sa]

Location of Poles and Zeros



ROOT LOCUS ON REAL AXIS



The root locus exists between (1-,-2) (-3,-4)



ASYMPTOTES AND CENTROID



BREAK-AWAY POINTS

$$KG(s)H(s) = \frac{K(s-3)(s-5)}{(s+1)(s+2)} = \frac{K(s^2-8s+15)}{(s^2+3s+2)}$$

Solving for K, and then differentiating, We get S = -1.45 and 3.82



BREAK – IN POINTS



Break-in point is found at -2.8 with k = 27.91

iω $\zeta = 0.8$ *j*4 -4.6 + j 3.45, K = 39.64j3 s-plane j2 -1.19 + j0.90, K = 12.79-0.87 + j0.66, K = 7.36• σ **X** -10 -9 -8 -7 -6 -5 -4-3 -1.5 -0 -j1 X = Closed-loop pole $\mathbf{X} = \text{Open-loop pole}$ -j2

ANGLE OF DEPARTURE and ARRIVAL jω K(s + 3)C(s)R(s) + j4 s(s+1)(s+2)(s+4)j3 s-plane Angle of j2 departure θ_1 *j*1 θ_3 θ_{A} ·σ $\theta_1 = -251.6^\circ = 108.4^\circ$ 0 -4-3 -2-1 θ_2 -j1

-j2

-j3

-j4

ANGLE OF ARRIVAL





 $-K^{2}-65K+720 = 0$, and solving for K= 9.65

CALCULATION OF GAIN VALUE K...

$(90 - K)s^{2} + 21K = 80.35s^{2} + 202.7 = 0$

<i>s</i> ⁴	1	14	3 <i>K</i>
<i>s</i> ³	7	8 + K	
<i>s</i> ²	90 - K	21 <i>K</i>	
s^1	$\frac{-K^2 - 65K + 720}{90 - K}$		
<i>s</i> ⁰	21 <i>K</i>		

S is found to be $\pm j1.59$. Thus the root locus crosses the imaginary axis at $\pm j1.59$ at a gain of 9.65. We conclude the system is stable for

 $0 \le K < 9.65$